

ECD-II
 solution. - ECD-II Page No. 4/6

Q. No.

Marks

➤ (a) LOG amplifier: Circuit diagram (2M)
 Derivation for output (2M)
 Disadvantages. (1M)

(b) Comparison - 5 points - (5M)
 Block diagram, A_f , R_{inf} , R_{of} , β ,
 definition.

(c) Full wave rectifier: Circuit diagram (2M)
 Working with input-output
 waveforms (3M)

2> (a) Colpitt's Oscillator: Circuit diagram (2M)
 Derivation (5M)
 Disadvantage (1M)
 Solution to the disadvantage: (2M)

(b) $P_L = 9\text{ W}$, $R_L = 6\Omega$, circuit (2M)

Assume $\eta_T = 0.9$, $P_L' = \frac{P_L}{\eta_T}$

$P_{d(max)} \geq 2P_L'$, select ECN149 (1M)

select V_{CC} ($V_{CC} < \frac{V_{CE0}}{2}$), $V_{RE} = 10\% V_{CC}$. (1/2M)

$$R_L' = \frac{(V_{CC} - V_{RE} - V_{CE(sat)})^2}{2P_L'} = \frac{V_{CE0}^2}{2P_L'} \quad (1M)$$

$$\frac{N_1}{N_2} = \sqrt{\frac{R_L'}{R_L}} \quad (1/2M)$$

$$I_{CQ} = \frac{2P_L'}{V_{CEQ}} \quad R_E = \frac{V_{RE}}{I_{EQ}} \quad (1M)$$

Assume $S \leq 10$, $S = \frac{(1 + \beta_{max})(1 + K)}{1 + \beta_{max} + K}$

$$\therefore K = 9.89 = \frac{R_{th}}{R_E} \quad (1M) \quad , \quad V_{th} = V_{BE} + I_C \left(\frac{R_{th}}{\beta_{typ}} + R_E + \frac{R_E}{\beta_{typ}} \right) \quad (1M)$$

Q. No.

$$R_1 = \frac{V_{CC} R_{th}}{V_{th}}$$

$$R_2 = \frac{V_{th} R_1}{V_{CC} - V_{th}} \quad (1M)$$

Marks

Find P_{R_1} and P_{R_2} and select R_1, R_2 . (1M)
 Select $C_C = 10 \mu F / V_{CC}$, $C_E = 100 \mu F / V_{CC}$.

Redraw circuit with all components. (1M)

Any other design steps may be considered.

4. (a) Shortcomings of ideal integrator:
 Can not be used at low frequency (1M)
 Practical integrator: Circuit diagram (2M)
 Derivation for gain (5M)
 Frequency response (2M)

5. (a) Class-B push-pull power amplifier:
 Circuit diagram (2M)
 Derivation for efficiency and figure of merit (5M)
 Advantages (1M)
 Disadvantages (1M)
 Output current waveform (1M)

- (c) Clamper: Circuit diagram (2M)
 working (2M)
 input - output waveforms (1M)

Q. No.

6. (a) Characteristics : 5 points (5M)
of -ve feedback
~~(6M)~~

(b) Class AB push-pull power amplifier : Circuit diagram (2M)
working (1M)
Input-output waveforms (1M)
Advantages (1M)

(c) Current mirror circuit (2M)
working (3M)

(d) Peak detector : Circuit (2M)
working (2M) Output waveform (1M)

(e) Schmitt trigger : Circuit (2M)
working with waveforms (3M)

Q. No.

Marks

Solution

Numericals.

$$1) (d) P_D (\text{without heat sink}) = \frac{T_{j_{\max}} - T_{\text{amb}}}{\theta_{\text{dev-case}} + \theta_{\text{case-amb}}}$$

$$P_D (\text{with heat sink}) = 1.86 \text{ W}$$

$$= \frac{T_{j_{\max}} - T_{\text{amb}}}{\theta_{\text{dev-case}} + \theta_{\text{case-sink}} + \theta_{\text{sink-amb}}}$$

$$= 16.1 \text{ W}$$

3) (a) Designing of instrumentation amplifier (4M)

$$A_f = \frac{R_3}{R_2} \left[1 + \frac{2R_f}{R_1} \right]$$

$$\text{Let } \frac{R_3}{R_2} = \frac{1}{5}$$

$$\text{Case (i)} \quad A_{f_{\min}} = 1 \quad \text{Case (ii)} \quad A_{f_{\max}} = 1000$$

$$\therefore 1 + \frac{2R_f}{R_{1_{\max}}} = 2$$

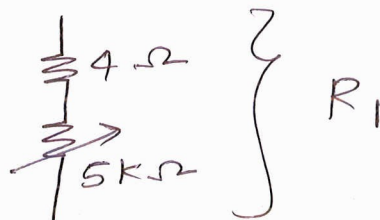
$$\therefore 1 + \frac{2R_f}{R_{1_{\min}}} = 5000$$

$$\therefore R_{1_{\max}} = \frac{R_f}{2}$$

$$\therefore R_{1_{\min}} = \frac{2R_f}{4999}$$

$$\text{Let } R_f = 10 \text{ k}\Omega \quad \therefore R_{1_{\min}} = 4 \Omega, R_{1_{\max}} = 5 \text{ k}\Omega$$

Let R_1 be a combination of one fixed resistor (4Ω) in series with variable resistor ($5 \text{ k}\Omega$)



Q. No.

$$3) (b) I_{BQ} = \frac{V_{EE} - V_{BE}}{R_S + 2\beta R_E} = 43 \mu A \quad (1M)$$

$$I_{EQ} = 6.5 mA \quad (1/2 M)$$

$$r_e = 4 \Omega \quad (1/2 M)$$

Derivation $-(4M)$

$$A_d = \frac{R_C}{2r_e} = 587.5 \quad (1/2 M)$$

$$A_c = \frac{-R_C}{2R_E} = -1.6 \quad (1/2 M)$$

$$CMRR = 20 \log_{10} \left| \frac{A_d}{A_c} \right| = 51.3 dB \quad (1M)$$

Need of swamping resistor $(2M)$

4) (b) Type: Voltage shunt negative feedback $(1M)$
 Circuit without feedback $(2M)$

$$R_M = \frac{-h_{fe} R'_S R'_C (1+h_{fe}) R_E}{(R_E + h_{ie}) (R'_S + h_{ie} + (1+h_{fe}) (R_E \parallel h_{ie}))} \quad (3M)$$

where $R'_S = R_S \parallel [R_F + (R_2 \parallel R_1)] = 990 \Omega$

$$R'_C = R_C \parallel [R_1 + (R_F \parallel R_2)] = 2.3 k\Omega$$

$$\therefore R_M = -219.067 k\Omega \quad (1M)$$

$$R_{in} \beta = \frac{-R_2}{R_F (R_1 + R_2)} = 4.76 \times 10^{-7} \Omega \quad (2M)$$

$$R_{in} = R'_S \parallel [h_{ie} + (1+h_{fe}) (R_E \parallel h_{ie})] \quad (2M)$$

$$= 971 \Omega$$

$$R_{Mf} = \frac{R_M}{1 + R_M \beta} = -198.38 k\Omega \quad (1M)$$

$$R_{inf} = \frac{R_{in}}{1 + R_{in} \beta} = 879.5 \Omega \quad (1M)$$

Q. No.

5) (b) Designing of RC phase shift oscillator

Circuit diagram - (2M)

Designing (3M)

$$f_{osc} = \frac{1}{2\pi RC\sqrt{6}}$$

Let $C = 0.01 \mu F$

$$\therefore R = 650 \Omega$$

$$\frac{R_f}{R} \geq 29$$

$$\therefore R_f \geq 19 k\Omega$$

Select $R_f = 21 k\Omega$

— X — X —